TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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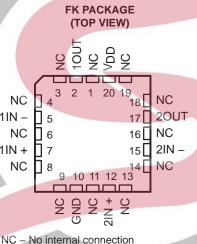
- Trimmed Offset Voltage: TLC27M7 . . . 500 μV Max at 25°C, V_{DD} = 5 V
- Input Offset Voltage Drift . . . Typically
 0.1 μV/Month, Including the First 30 Days
- Wide Range of Supply Voltages Over Specified Temperature Ranges:

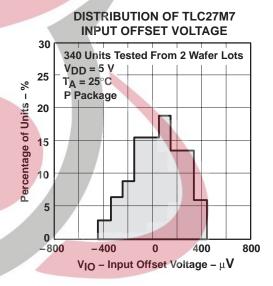
0°C to 70°C . . . 3 V to 16 V -40°C to 85°C . . . 4 V to 16 V -55°C to 125°C . . . 4 V to 16 V

- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)

- Low Noise . . . Typically 32 nV/√Hz at f = 1 kHz
- Low Power . . . Typically 2.1 mW at 25°C,
 V_{DD} = 5 V
- Output Voltage Range Includes Negative Rail
- High Input impedance . . . $10^{12} \Omega$ Typ
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latch-Up Immunity

D, JG, P OR PW PACKAGE (TOP VIEW) 10UT Vcc 20UT 1IN -2 6 12IN -1IN + 3 GND 5 2IN + NC 1IN NC 1IN + NC





AVAILABLE OPTIONS

				PACKAGE		
TA	V _{IO} max AT 25°C	SMALL OUTLINE	CHIP CARRIER	CERAMIC DIP	PLASTIC DIP	TSSOP
		(D)	(FK)	(JG)	(P)	(PW)
	500 μV	TLC27M7CD		_	TLC27M7CP	
000 to 7000	2 mV	TLC27M2BCD	_	_	TLC27M2BCP	_
0°C to 70°C	5 mV	TLC27M2ACD		_ = _	TLC27M2ACP	_
	10 mV	TLC27M2CD	E + E	CTR	TLC27M2CP	TLC27M2CPW
	500 μV	TLC27M7ID			TLC27M7IP	_
4000 1- 0500	2 mV	TLC27M2BID	_	_	TLC27M2BIP	_
-40°C to 85°C	5 mV	TLC27M2AID	_	_	TLC27M2AIP	_
	10 mV	TLC27M2ID	_	_	TLC27M2IP	TLC27M2IPW
−55°C to 125°C	500 μV	TLC27M7MD	TLC27M7MFK	TLC27M7MJG	TLC27M7MP	_
-55°C to 125°C	10 mV	TLC27M2MD	TLC27M2MFK	TLC27M2MJG	TLC27M2MP	_

The D and PW package are available taped and reeled. Add R suffix to the device type (e.g., TLC27M7CDR). For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

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ts de

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SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

description

The TLC27M2 and TLC27M7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M2 (10 mV) to the high-precision TLC27M7 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M2 and TLC27M7. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

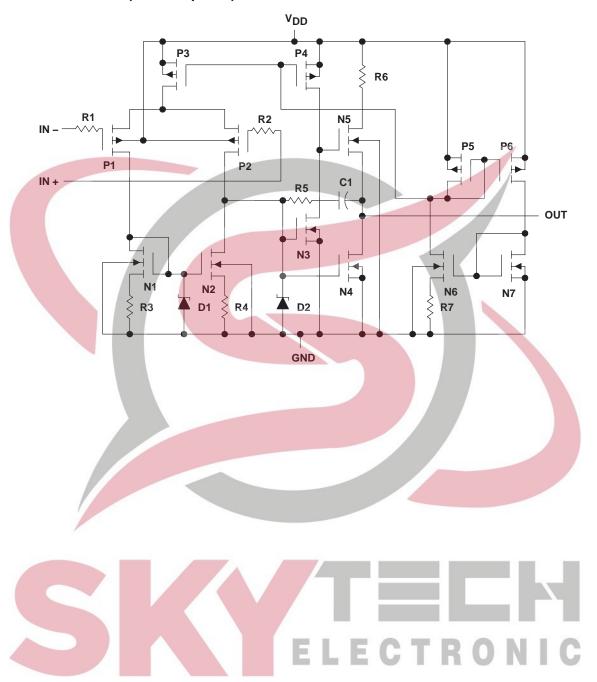
The device inputs and outputs are designed to withstand -100-mA surge currents without sustaining latch-up.

The TLC27M2 and TLC27M7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C.



equivalent schematic (each amplifier)



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SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V _{DD} (see Note 1)	
Differential input voltage, V _{ID} (see Note 2)	±V _{DD}
Input voltage range, V _I (any input)	
Input current, I ₁	±5 mA
Output current, I _O (each output)	±30 mA
Total current into V _{DD}	45 mA
Total current out of GND	
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T _A : C suffix	0°C to 70°C
Operating free-air temperature, T _A : C suffix	
Operating free-air temperature, T _A : C suffix	
Operating free-air temperature, T _A : C suffix I suffix M suffix Storage temperature range	
Operating free-air temperature, T _A : C suffix I suffix M suffix Storage temperature range Case temperature for 60 seconds: FK package	
Operating free-air temperature, T _A : C suffix I suffix M suffix Storage temperature range	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
 - 2. Differential voltages are at IN+ with respect to IN-.
 - 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
Р	1000 mW	8.0 mW/°C	640 mW	520 mW	

recommended operating conditions

		_	C SUI	FFIX	I SUI	FIX	M SU	FFIX	LINUT
			MIN	MAX	MIN	MAX	MIN	MAX	UNIT
Supply voltage, V _{DD}			3	16	4	16	4	16	V
	V _{DD} = 5 V		-0.2	3.5	-0.2	3.5	0	3.5	.,
Common-mode input voltage, V _{IC}	V _{DD} = 10 V		-0.2	8.5	-0.2	8.5	0	8.5	V
Operating free-air temperature, TA			0	70	-40	85	-55	125	_ °C

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SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

	PARAMETER		TEST CONE	DITIONS	T _A †	TL TL TL	.C27M2 .C27M2 .C27M2 .C27M7	AC BC C	UNIT
		1			25°C	MIN	TYP 1.1	MAX	
		TLC27M2C	$V_{O} = 1.4 \text{ V},$ $R_{S} = 50 \Omega,$	$V_{IC} = 0$, $R_{I} = 100 \text{ k}\Omega$	Full range		1.1	10 12	
			-	•	25°C		0.9	5	mV
		TLC27M2AC	$V_{O} = 1.4 \text{ V},$ $R_{S} = 50 \Omega,$	$V_{IC} = 0$, $R_I = 100 \text{ k}\Omega$	Full range		0.0	6.5	
VIO	Input offset voltage		$V_{O} = 1.4 \text{ V},$	V _{IC} = 0,	25°C		220	2000	
		TLC27M2BC	$R_S = 50 \Omega$	$R_{\parallel} = 100 \text{ k}\Omega$	Full range	71		3000	
			V _O = 1.4 V,	V _{IC} = 0,	25°C		185	500	μV
		TLC27M7C	$R_S = 50 \Omega$,	$R_{\rm I} = 100 \text{ k}\Omega$	Full range			1500	
αΝΙΟ	Average temperature coeffi offset voltage	cient of input			25°C to 70°C		1.7		μV/°C
					25°C		0.1	60	
lio	Input offset current (see No	te 4)	$V_0 = 2.5 \text{ V},$	$V_{IC} = 2.5 V$	70°C		7	300	рA
		4)	0.51/	.,	25°C		0.6	60	
Iв	Input bias current (see Note	9 4)	$V_{O} = 2.5 \text{ V},$	V _{IC} = 2.5 V	70°C		40	600	pA
	Common-mode input voltage	ge range		-	25°C	-0.2 to 4	-0.3 to 4.2		٧
VICR	(see Note 5)				Full range	-0.2 to 3.5			٧
					25°C	3.2	3.9		
Vон	High-level output voltage		V _{ID} = 100 mV,	$R_L = 100 \text{ k}\Omega$	0°C	3	3.9		V
					70°C	3	4		
					25°C		0	50	
V_{OL}	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	I _{OL} = 0	0°C		0	50	mV
					70°C		0	50	
	Large-signal differential vol	ane			25°C	25	170		
AVD	amplification	age	$V_0 = 0.25 \text{ V to 2 V},$	$R_L = 100 \text{ k}\Omega$	0°C	15	200		V/mV
					70°C	15	140		
					25°C	65	91		
CMRR	Common-mode rejection ra	tio	VIC = VICRmin		0°C	60	91		dB
					70°C	60	92		
kSVR	Supply-voltage rejection rat (ΔV _{DD} /ΔV _{IO})	io	V _{DD} = 5 V to 10 V,	V _O = 1.4 V	25°C 0°C	70 60	93 92	C	dB
	(AVDD/AVIO)				70°C	60	94]
	-		0.517		25°C		210	560	
I_{DD}	Supply current (two amplifie	ers)	V _O = 2.5 V, No load	$V_{IC} = 2.5 V$,	0°C		250	640	μΑ
	i- 000 t- 7000				70°C		170	440	

[†] Full range is 0°C to 70°C.



^{5.} This range also applies to each input individually.

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SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

	PARAMETER		TEST COM	IDITIONS	T _A †	TL TL	C27M20 C27M2/ C27M2I C27M70	AC BC	UNIT
						MIN	TYP	MAX	
		TLC27M2C	V _O = 1.4 V,	$V_{IC} = 0$,	25°C		1.1	10	
		TEGETIVIZE	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			12	mV
		TLC27M2AC	$V_{O} = 1.4 V$,	$V_{IC} = 0$,	25°C		0.9	5	1111
VIO	Input offset voltage	TEGETWIZAG	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			6.5	
100	input onset voltage	TLC27M2BC	$V_0 = 1.4 V$	$V_{IC} = 0$,	25°C		224	2000	
		TEOZIWIZBO	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			3000	μV
		TLC27M7C	$V_0 = 1.4 V$	$V_{IC} = 0$,	25°C		190	800	μν
		TEOZIWI	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			1900	
ανιο	Average temperature coe offset voltage	fficient of input			25°C to 70°C		2.1		μV/°C
	Oliset voltage				25°C		0.1	60	
IIO	Input offset current (see N	lote 4)	$V_0 = 5 V$	V _{IC} = 5 V	70°C		7	300	pА
					25°C		0.7	60	
I _{IB}	Input bias current (see No	ote 4)	V _O = 5 V,	$V_{IC} = 5 V$	70°C		50	600	pА
					70°C	-0.2		600	
	Common-mode input volta	age range			25°C	to 9	-0.3 to 9.2		V
VICR	(see Note 5)				Full range	-0.2 to 8.5			V
					25°C	8	8.7		
Vон	High-level output voltage		$V_{ID} = 100 \text{ mV},$	$R_L = 100 \text{ k}\Omega$	0°C	7.8	8.7		V
					70°C	7.8	8.7		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	I _{OL} = 0	0°C		0	50	mV
					70°C		0	50	
	Lorge signal differential w	olto a o			25°C	25	275		
AVD	Large-signal differential vo amplification	oitage	$V_0 = 1 V to 6 V,$	$R_L = 100 \text{ k}\Omega$	0°C	15	320		V/mV
					70°C	15	230		
					25°C	65	94		
CMRR	Common-mode rejection	ratio	V _{IC} = V _{ICR} min		0°C	60	94		dB
					70°C	60	94		
	Supply-voltage rejection r	atio		FIE	25°C	70	93		
ksvr	(ΔV _{DD} /ΔV _{IO})	allo	$V_{DD} = 5 \text{ V to } 10 \text{ V,}$	$V_0 = 1.4 \text{ V}$	0°C	60	92		∅ dB
					70°C	60	94		
			V _O = 5 V,	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	25°C		285	600	
IDD	Supply current (two ampli	fiers)	VO = 5 V, No load	$V_{IC} = 5 V$	0°C		345	800	μΑ
					70°C		220	560	

[†] Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

Vio		PARAMETER		TEST CON	DITIONS	T _A †	TL TL TL	.C27M2l .C27M2l .C27M2l .C27M7l	AI BI	UNIT
Vio Input offset voltage TLC27M2I RS = 50 Ω, RL = 100 kΩ Full range 13 TLC27M2AI RS = 50 Ω, RL = 100 kΩ Full range 77 TLC27M2AI RS = 50 Ω, RL = 100 kΩ Full range 78 TLC27M2AI RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M2BI RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M2BI RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M7I RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M7I RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M7I RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M7I RS = 50 Ω, RL = 100 kΩ Full range 79 TLC27M7I TLC27M7I RS = 50 Ω, RL = 100 kΩ TLC27M2BI TLC27M7I TLC27M7I RS = 50 Ω, RL = 100 kΩ TLC27M7I			I			27:0	MIN	TYP	MAX	
Vio Input offset voltage TLC27M2AI VO = 1.4 V, RS = 50 Ω, RL = 100 kΩ Full range TO TLC27M2AI RS = 50 Ω, RL = 100 kΩ Full range TO TLC27M2AI RS = 50 Ω, RL = 100 kΩ Tull range TO TLC27M2BI RS = 50 Ω, RL = 100 kΩ Full range To Tull range Tull ra			TLC27M2I	$V_0 = 1.4 \text{ V},$	$V_{IC} = 0$,			1.1		
Vo					_			0.0		mV
Vo			TLC27M2AI		V _{IC} = 0,			0.9		
TLC27M2BI No = 1.4 V, No = 0. RL = 100 kΩ Full range 3500 μ V RS = 50 Ω, RL = 100 kΩ Full range 3500 μ V RS = 50 Ω, RL = 100 kΩ Full range 25°C 185 500 μ V RS = 50 Ω, RL = 100 kΩ Full range 2000	VIO	Input offset voltage		-			1			
TLC27M7I V _O = 1.4 V _I V _{IC} = 0, R _L = 100 kΩ Full range 2000 Full range 2000 R _L = 100 kΩ Full range 25°C 1.7 μV/°C 100 R _L = 100 kΩ R _L	.0		TLC27M2BI		$V_{IC} = 0,$			220		
TLC27M7I V _O = 1.4 V, V _{IC} = 0.				-						μV
Average temperature coefficient of input offset voltage 1.7 μV/°C 25°C 0.1 60 85°C 24 1000 pA	/		TLC27M7I			_		185		
Indicate				$RS = 50 \Omega$	R _L = 100 kΩ				2000	
Input offset current (see Note 4)	ανιο		icient of input					1.7		μV/°C
I I I I I I I I I I		Land offert compatition N	. A	V 05.V	V 05V	25°C		0.1	60	A
Input bias current (see Note 4) V _O = 2.5 V, V _{IC} = 2.5 V 85°C 200 2000 pA	IO	input onset current (see No	ote 4)	$V_0 = 2.5 \text{ V},$	VIC = 2.5 V	85°C		24	1000	рА
Vicro Common-mode input voltage range (see Note 5) V Vicro					.,	25°C		0.6	60	
Vocation	ΙΒ	Input bias current (see Not	e 4)	$V_0 = 2.5 \text{ V},$	$V_{IC} = 2.5 \text{ V}$	85°C		200	2000	рA
VICR (see Note 5) Full range to 3.5 -0.2 to 3.5 V VOH High-level output voltage VID = 100 mV, RL = 100 kΩ -40° C 3.2.3.9 V VOL Low-level output voltage VID = -100 mV, IOL = 0 -40° C 0.50 mV AVD Large-signal differential voltage amplification VO = 0.25 V to 2 V, RL = 100 kΩ -40° C 15.270 V/mV CMRR Common-mode rejection ratio VIC = VICRmin -40° C 60.90 90 dB kSVR Supply-voltage rejection ratio (AVDD/ΔVIO) VDD = 5 V to 10 V, VO = 1.4 V -40° C 60.91 91 dB IDD Supply-voltage rejection ratio (two amplifiers) VO = 2.5 V, VIC = 2.5 V, VIC = 2.5 V, VIC = 2.5 V, OVIC		Common-mode input volta	ge range		-	25°C	to	to		V
$V_{OH} \text{High-level output voltage} \qquad V_{ID} = 100 \text{mV}, \qquad R_L = 100 \text{k}\Omega \qquad -40^{\circ}\text{C} \qquad 3 3.9 \qquad V$ $V_{OL} \text{Low-level output voltage} \qquad V_{ID} = -100 \text{mV}, \qquad I_{OL} = 0 \qquad 25^{\circ}\text{C} \qquad 0 \qquad 50 \qquad \text{mV}$ $A_{VD} \text{Large-signal differential voltage amplification} \qquad V_{O} = 0.25 \text{V to 2 V}, R_L = 100 \text{k}\Omega \qquad -40^{\circ}\text{C} \qquad 15 \qquad 270 \qquad \text{V/mV}$ $CMRR \text{Common-mode rejection ratio} \qquad V_{IC} = V_{ICR} \text{min} \qquad -40^{\circ}\text{C} \qquad 60 90 \qquad \text{dB}$ $k_{SVR} \text{Supply-voltage rejection ratio} \qquad V_{DD} = 5 \text{V to 10 V}, V_{O} = 1.4 \text{V}$ $V_{DD} = 5 \text{V to 10 V}, V_{O} = 1.4 \text{V}$ $V_{DD} = 2.5 \text{V}, V_{IC} =$	VICR	•				Full range	to			٧
$V_{OL} \text{Low-level output voltage} \qquad V_{ID} = -100 \text{mV}, I_{OL} = 0 \qquad \frac{25^{\circ}\text{C}}{85^{\circ}\text{C}} \qquad 0 \qquad 50 \\ \hline & & & & & & & & & & & & & & & & & &$						25°C	3.2	3.9		
V _{OL} Low-level output voltage $V_{ID} = -100 \text{ mV}, I_{OL} = 0$ $A_{VD} = -100 $	Vон	High-level output voltage		V _{ID} = 100 mV,	$R_L = 100 \text{ k}\Omega$	-40°C	3	3.9		V
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$						85°C	3	4	1	
AVD Large-signal differential voltage amplification $V_{O} = 0.25 \text{ V to 2 V}, R_{L} = 100 \text{ k}Ω$ 25°C 25 170 25°C 25 170 25°C 25 170 25°C 25°C 25 170 25°C 25°						25°C		0	50	
$AVD \text{Large-signal differential voltage amplification} V_O = 0.25 \text{ V to 2 V}, R_L = 100 \text{ k}\Omega \qquad \frac{25^\circ\text{C}}{-40^\circ\text{C}} \qquad \frac{25}{15} \qquad \frac{170}{130} \qquad V/\text{mV}$ $\frac{85^\circ\text{C}}{-40^\circ\text{C}} \qquad \frac{15}{130} \qquad \frac{270}{15} \qquad \frac{170}{130} \qquad \frac{170}$	VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	I _{OL} = 0	-40°C		0	50	mV
AVD Large-signal differential voltage amplification $V_O = 0.25 \text{ V to 2 V}$, $R_L = 100 \text{ kΩ}$ -40°C 15 270 V/mV CMRR Common-mode rejection ratio VIC = VICRmin -40°C 66 91 CMRR Supply-voltage rejection ratio $V_{OD} = 5 \text{ V to 10 V}$, $V_{O} = 1.4 \text{ V}$ -40°C 60 91 60 Large-signal differential voltage 85°C 65 91 $V_{OD} = 5 \text{ V to 10 V}$, $V_{O} = 1.4 \text{ V}$ -40°C 60 91 $V_{OD} = 5 \text{ V to 10 V}$, $V_{O} = 1.4 \text{ V}$ -40°C 60 91 $V_{OD} = 2.5 \text{ V}$, $V_{O} = 2.5 $						85°C		0	50	
AVD amplification						25°C	25	170		
	AVD		ltage	$V_0 = 0.25 \text{ V to 2 V},$	$R_L = 100 \text{ k}\Omega$	-40°C	15	270		V/mV
CMRR Common-mode rejection ratio $V_{IC} = V_{ICR}$ min $-40^{\circ}C$ 60 90 dB $85^{\circ}C$ 60 90 dB $85^{\circ}C$ 60 90 dB $V_{DD} = 5 \text{V to } 10 \text{ V}, V_{O} = 1.4 \text{ V}$ $-40^{\circ}C$ 60 91 00 00 00 00 00 00 00 0		amplification				85°C	15	130		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			/ / /			25°C	65	91		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	CMRR	Common-mode rejection ra	atio	VIC = VICRMIN		-40°C	60	90		dB
$V_{DD} = 5$ V to 10 V, $V_{O} = 1.4$ V $V_{$				10 1010		85°C	60	90		
$V_{DD} = 5$ V to 10 V, $V_{O} = 1.4$ V $V_{$					FA	25°C	70	93	A	
85°C 60 94 VO = 2.5 V, VIC = 2.5 V, VIC = 2.5 V, 315 800 UA	ks//p		itio	VDD = 5 V to 10 V.	Vo = 1.4 V				U	dB
V _O = 2.5 V, V _{IC} = 2.5 V, V _{IC} = 2.5 V, 25°C 210 560 V _O = 2.5 V, V _{IC} = 2.5 V, 240°C 315 800 V _{IC}	501	(ΔΛDD/ΔΛΙΟ)		,	0		60	94		
Supply current (two amplifiers) $V_0 = 2.5 \text{ V}, V_{1C} = 2.5 \text{ V}, = 40^{\circ}\text{C}$		Annual Annual							560	
I NO load	IDD	Supply current (two amplifi	ers)		$V_{IC} = 2.5 V$,					μΑ
85°C 160 400		11.5	,	No load						.

[†] Full range is –40°C to 85°C.



^{5.} This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

	PARAMETER		TEST CONI	DITIONS	TAT	TI TI	LC27M2 LC27M2 LC27M2 LC27M7 TYP	AI BI	UNIT
			V _O = 1.4 V,	V _{IC} = 0,	25°C	101111	1.1	10	
		TLC27M2I	$R_S = 50 \Omega$	$R_L = 100 \text{ k}\Omega$	Full range			13	
			V _O = 1.4 V,	V _{IC} = 0,	25°C		0.9	5	mV
		TLC27M2AI	$R_S = 50 \Omega$	$R_L = 100 \text{ k}\Omega$	Full range		- /	7	
VIO	Input offset voltage		V _O = 1.4 V,	$V_{IC} = 0$,	25°C		224	2000	
		TLC27M2BI	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range		7/	3500	
		TI 0071471	V _O = 1.4 V,	V _{IC} = 0,	25°C		190	800	μV
		TLC27M7I	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			2900	
ανιο	Average temperature coeffice offset voltage	cient of input			25°C to 85°C	7	2.1		μV/°C
	land offers and the Alexander		V 5V	V 5.V	25°C		0.1	60	4
lio	Input offset current (see No	te 4)	$V_0 = 5 V$,	$V_{IC} = 5 V$	85°C		26	1000	рA
					25°C		0.7	60	
I _{IB}	Input bias current (see Note	: 4)	V _O = 5 V,	V _{IC} = 5 V	85°C		220	200 0	pА
\\.	Common-mode input voltag	e range			25°C	-0.2 to 9	-0.3 to 9.2		V
VICR	(see Note 5)				Full range	-0.2 to 8.5			٧
					25°C	8	8.7		
Vон	High-level output voltage		$V_{ID} = 100 \text{ mV},$	$R_L = 100 \text{ k}\Omega$	-40°C	7.8	8.7		V
					85°C	7.8	8.7		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	-40°C		0	50	mV
					85°C		0	50	
	Large signal differential valt	000			25°C	25	275		
A_{VD}	Large-signal differential volt amplification	age	$V_0 = 1 \text{ V to 6 V},$	R _L = 100 kΩ	-40°C	15	390		V/mV
					85°C	15	220	7	
					25°C	65	94		
CMRR	Common-mode rejection ra	tio	$V_{IC} = V_{ICR}min$		-40°C	60	93		dB
					85°C	60	94		
	Supply-voltage rejection rati	io			25°C	70	93		-
ksvr	Supply-voltage rejection rati (ΔVDD/ΔVIO)	10	$V_{DD} = 5 \text{ V to } 10 \text{ V},$	$V_0 = 1.4 \text{ V}$	-40°C	60	91		dB
	. 55 10/				85°C	60	94		
			Vo - 5 V	Vio – F.V	25°C		285	600	
IDD	Supply current		V _O = 5 V, No load	$V_{IC} = 5 V$,	-40°C		450	900	μΑ
<u> </u>	an in 4000 to 9500				85°C		205	520	

[†] Full range is -40° C to 85° C.

^{5.} This range also applies to each input individually.



TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, $V_{DD} = 5 \text{ V}$ (unless otherwise noted)

	PARAMETER		TEST CON	DITIONS	T _A †		.C27M2l		UNIT
						MIN	TYP	MAX	
		TI 00714014	V _O = 1.4 V,	V _{IC} = 0,	25°C		1.1	10	
. ,	Lamest affact welliams	TLC27M2M	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			12	
VIO	Input offset voltage	TI 00714714	V _O = 1.4 V,	V _{IC} = 0,	25°C		185	500	mV
		TLC27M7M	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			3750	
ανιο	Average temperature coeff offset voltage	icient of input			25°C to 125°C		1.7		μV/°C
	Land offers and the Alle	(1-4)	V 05V	V 05V	25°C		0.1	60	pA
lio	Input offset current (see No	ote 4)	$V_0 = 2.5 \text{ V},$	$V_{IC} = 2.5 V$	125°C		1.4	15	nA
. /	1 (1) (4 N)		V 05V	V 0.5W	25°C		0.6	60	pА
IB	Input bias current (see Not	e 4)	$V_0 = 2.5 V$,	$V_{IC} = 2.5 V$	125°C	7	9	35	nA
						0	-0.3		
\ \					25°C	to	to		V
VICR	Common-mode input voltage	ge range			_	4	4.2		
lok	(see Note 5)				Full range	0 to			V
					T dir range	3.5			V
					25°C	3.2	3.9		
Vон	High-level output voltage		V _{ID} = 100 mV,	$R_L = 100 \text{ k}\Omega$	−55°C	3	3.9		V
0.1			10	-	125°C	3	4		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	-55°C		0	50	mV
0	3		ID ,	OL 1	125°C		0	50	
					25°C	25	170		
AVD	Large-signal differential vol	tage	$V_0 = 0.25 \text{ V to 2 V},$	$R_{\rm I} = 100 \rm k\Omega$	−55°C	15	290		V/mV
"	amplification			_	125°C	15	120	/	
					25°C	65	91		
CMRR	Common-mode rejection ra	atio	V _{IC} = V _{ICR} min		−55°C	60	89		dB
			10 101		125°C	60	91		
					25°C	70	93		
ksvr	Supply-voltage rejection ra	tio	$V_{DD} = 5 \text{ V to } 10 \text{ V},$	V _O = 1.4 V	-55°C	60	91		dB
	$(\Delta V_{DD}/\Delta V_{IO})$				125°C	60	94		
					25°C		210	560	
I _{DD}	Supply current (two amplifications)	ers)	$V_0 = 2.5 \text{ V},$	$V_{IC} = 2.5 V,$	-55°C		340	880	μА
			No load		125°C		_ 140	360	
<u> </u>									

[†] Full range is -55°C to 125°C.

^{5.} This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

	PARAMETER		TEST CON	DITIONS	T _A †		C27M2N C27M7N		UNIT
						MIN	TYP	MAX	
		TI 00714014	V _O = 1.4 V,	V _{IC} = 0,	25°C		1.1	10	
.,		TLC27M2M	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			12	.,
VIO	Input offset voltage	TI 00714714	V _O = 1.4 V,	V _{IC} = 0,	25°C		190	800	mV
		TLC27M7M	$R_S = 50 \Omega$,	$R_L = 100 \text{ k}\Omega$	Full range			4300	
αΛΙΟ	Average temperature coeffice offset voltage	eient of input			25°C to 125°C		2.1		μV/°C
					25°C		0.1	60	
lΟ	Input offset current (see Not	e 4)	$V_0 = 5 V$,	$V_{IC} = 5 V$	125°C		1.8	15	рA
					25°C		0.7	60	
ΙΒ	Input bias current (see Note	4)	$V_0 = 5 V$,	$V_{IC} = 5 V$	125°C		10	35	рA
V	Common-mode input voltage	e range			25°C	0 to 9	-0.3 to 9.2		V
VICR	(see Note 5)				Full range	0 to 8.5			V
					25°C	8	8.7		
V_{OH}	High-level output voltage		$V_{ID} = 100 \text{ mV},$	$R_L = 100 \text{ k}\Omega$	−55°C	7.8	8.6		V
					125°C	7.8	8.8		
					25°C		0	50	
V_{OL}	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	-55°C		0	50	mV
					125°C		0	50	
					25°C	25	275		
AVD	Large-signal differential volta amplification	age	$V_0 = 1 \text{ V to 6 V},$	$R_L = 100 \text{ k}\Omega$	−55°C	15	420		V/mV
	amplification	The second second			125°C	15	190	7	
					25°C	65	94		
CMRR	Common-mode rejection rat	io	VIC = VICRmin		−55°C	60	93		dB
					125°C	60	93		
					25°C	70	93		
ksvr	Supply-voltage rejection ration (ΔV _{DD} /ΔV _{IO})	0	$V_{DD} = 5 \text{ V to } 10 \text{ V},$	V _O = 1.4 V	-55°C	60	91		dB
	(A v DD/A v O)				125°C	60	94	7	
					25°C		285	600	
I_{DD}	Supply current (two amplifie	rs)	$V_O = 5 V$	$V_{IC} = 5 V$	-55°C		490	1000	μΑ
			No load						l

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

operating characteristics at specified free-air temperature, $V_{DD} = 5 \text{ V}$

	PARAMETER	TEST CO	ONDITIONS	TA	TLC27M TLC27M TLC27M TLC27M	2AC 2BC	UNIT
					MIN TY	P MAX	
				25°C	0.4	3	
			V _{I(PP)} = 1 V	0°C	0.4	6	
CD.	Class nata at switz main	$R_L = 100 \text{ k}\Omega$		70°C	0.3	6	\//
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C	0.4	0	V/μs
		3	$V_{I(PP)} = 2.5 \text{ V}$	0°C	0.4	3	
				70°C	0.3	4	
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C	3	2	nV/√ Hz
				25°C	5	5	
ВОМ	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_I = 100 \text{ k}\Omega$,		0°C	6	0	kHz
			See Figure 1	70°C	5	0	
				25°C	52	5	
В1	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 pF$,	0°C	60	0	kHz
		occ rigare o		70°C	40	0	
		.,		25°C	40	0	
φm	Phase margin	$V_{\parallel} = 10 \text{ mV},$ $C_{\parallel} = 20 \text{ pF},$		0°C	41	0	
		OL = 20 pr ,	_ ccc r iguic o	70°C	39	0	

operating characteristics at specified free-air temperature, V_{DD} = 10 V

	PARAMETER	TEST CO	ONDITIONS	TA	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C MIN TYP MAX	UNIT
				25°C	0.62	
			V _{I(PP)} = 1 V	0°C	0.67	1
	,	$R_L = 100 \text{ k}\Omega$,		70°C	0.51	
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C	0.56	V/μs
		Goo'l igulo 1	V _{I(PP)} = 5.5 V	0°C	0.61	1
				70°C	0.46]
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C	32	nV/√ Hz
			LEC.	25°C	35	
ВОМ	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100 \text{ k}\Omega$,	C _L = 20 pF, See Figure 1	0°C	40	kHz
			See rigure r	70°C	30]
		.,,		25°C	635	
В1	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 pF$,	0°C	710	kHz
		Gee rigare o		70°C	510	
		V 40 V	, ,	25°C	43°	
φm	Phase margin	$V_{I} = 10 \text{ mV},$ $C_{L} = 20 \text{ pF},$	f = B ₁ , See Figure 3	0°C	44°	
		2 2 pr,		70°C	42°	

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

operating characteristics at specified free-air temperature, V_{DD} = 5 V

	PARAMETER	TEST CO	ONDITIONS	TA	TLC27I TLC27I TLC27I TLC27I	UNIT		
					MIN TY	P MAX		
				25°C	0.4	3		
			V _{I(PP)} = 1 V	−40°C	0.5	1		
	Olements of male	$R_L = 100 \text{ k}\Omega$		85°C	0.3	5] ,,, ,	
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C	0.4)	V/μs	
		ord Higher	V _{I(PP)} = 2.5 V	-40°C	0.4	3]	
			1	85°C	0.3	2]	
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C	3	2	nV/√ Hz	
				25°C	5	5	kHz	
Вом	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_I = 100 \text{ k}\Omega$,		-40°C	7	5		
			See Figure 1	85°C	4	5		
				25°C	52	5		
В1	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 pF$,	-40°C	77)	kHz	
		Gee rigule 3		85°C	37)		
				25°C	40	0		
φm	Phase margin	$V_{\parallel} = 10 \text{ mV},$ $C_{\parallel} = 20 \text{ pF},$	f = B ₁ , See Figure 3	-40°C	43	0		
		OL - 20 pr ,	CCC i iguic 9	85°C	38	0		

operating characteristics at specified free-air temperature, $V_{DD} = 10 \text{ V}$

	PARAMETER	TEST CO	ONDITIONS	ТА	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I	UNIT
				25°C	MIN TYP MAX	
			V _{I(PP)} = 1 V	-40°C	0.77	1
		$R_L = 100 \text{ k}\Omega$,	1 ((PP)	85°C	0.47	1
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C	0.56	V/μs
		occ rigure r	V _{I(PP)} = 5.5 V	-40°C	0.70	1
				85°C	0.44	
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C	32	nV/√Hz
	Maximum output-swing bandwidth		5 L. F (25°C	35	
Вом		$V_O = V_{OH}$, $R_L = 100 \text{ k}\Omega$,	C _L = 20 pF, See Figure 1	_40°C	45	kHz
			occ riguio r	85°C	25	
		., ,,	0 00 5	25°C	635	
B ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 pF$,	−40°C	880	kHz
		Goo'r igai'o o		85°C	480	
		\\. 40\\	. D	25°C	43°	
φm	Phase margin	$V_{l} = 10 \text{ mV},$ $C_{L} = 20 \text{ pF},$	f = B ₁ , See Figure 3	-40°C	46°]
				85°C	41°	

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051E - OCTOBER 1987 - REVISED AUGUST 2008

operating characteristics at specified free-air temperature, $V_{DD} = 5 V$

	PARAMETER	TEST CO	ONDITIONS	TA	TL TL	UNIT		
					MIN	TYP	MAX	
				25°C		0.43		
			V _{I(PP)} = 1 V	−55°C		0.54		
		$R_L = 100 \text{ k}\Omega$		125°C		0.29		\ ,,,
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		0.40		V/μs
		$V_{I(PP)} = 2.5 \text{ V}$ -55°C 125°C	0.49					
				125°C		0.28		
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C		32		nV/√ Hz
		$V_O = V_{OH}$, $R_L = 100 \text{ k}\Omega$,	C _L = 20 pF,	25°C		55		
ВОМ	Maximum output-swing bandwidth			−55°C	7	80		kHz
			See rigure r	125°C		40		
			and the second	25°C		525		
В1	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 \text{ pF},$	-55°C		850		kHz
		occ r igure o		125°C		330		
		V 15 V		25°C		40°		
φm	Phase margin	$V_{\parallel} = 10 \text{ mV},$ $C_{\parallel} = 20 \text{ pF},$	$t = B_1$, See Figure 3	−55°C		44°		
		OL - 20 pr ,	CCC Figure 3	125°C		36°		

operating characteristics at specified free-air temperature, V_{DD} = 10 V

	PARAMETER	TEST CO	ONDITIONS	TA	TL TL	UNIT			
					MIN	TYP	MAX		
				25°C		0.62			
			V _{I(PP)} = 1 V	−55°C		0.81			
0.0	Olements at All Marks	$R_L = 100 \text{ k}\Omega$		125°C		0.38] ,,, ₋	
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		0.56		V/μs	
		gare :	V _{I(PP)} = 5.5 V	−55°C		0.73			
				125°C		0.35			
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20 \Omega$,	25°C		32		nV/√ Hz	
	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100 \text{ k}\Omega$,	C _L = 20 pF, See Figure 1	25°C		35		kHz	
ВОМ				−55°C		50			
				125°C		20			
		. E.	E.C.	25°C		635			
В1	Unity gain bandwidth	V _I = 10 mV, See Figure 3	$C_L = 20 pF$,	−55°C		960		kHz	
		Gee rigure 3		125°C		440			
	_		, _	25°C		43°			
φm	Phase margin	$V_{ } = 10 \text{ mV},$	f = B ₁ , See Figure 3	−55°C		47°			
		OL – 20 pr.,	oce i iguie 3	125°C		39°			

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27M2 and TLC27M7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

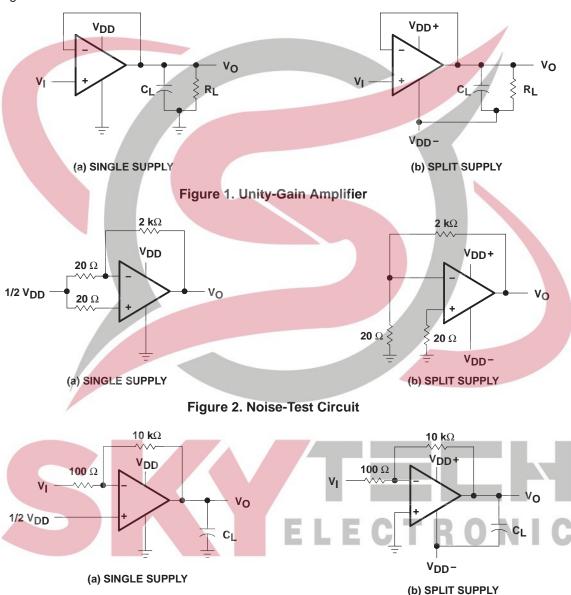


Figure 3. Gain-of-100 Inverting Amplifier

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC27M2 and TLC27M7 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

- 1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
- 2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution—many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

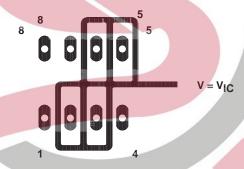


Figure 4. Isolation Metal Around Device Inputs (JG and P packages)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

PARAMETER MEASUREMENT INFORMATION

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage, since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways; full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

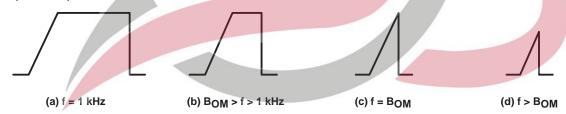


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

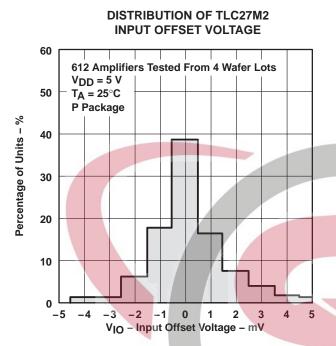
TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V _{IO}	Input offset voltage	Distribution	6, 7
α VIO	Temperature coefficient	Distribution	8, 9
Vон	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	10, 11 12 13
VOL	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	14, 15 16 17 18, 19
AVD	Differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	20 21 32, 33
I _{IB} /I _{IO}	Input bias and input offset current	vs Free-air temperature	22
V _{IC}	Common-mode input voltage	vs Supply voltage	23
I _{DD}	Supply current	vs Supply voltage vs Free-air temperature	24 25
SR	Slew rate	vs Supply voltage vs Free-air temperature	26 27
	Normalized slew rate	vs Free-air temperature	28
VO(PP)	Maximum peak-to-peak output voltage	vs Frequency	29
В ₁	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	30 31
^ф т	Phase margin	vs Supply voltage vs Free-air temperature vs Capacitive loads	34 35 36
V _n	Equivalent input noise voltage	vs Frequency	37
ф	Phase shift	vs Frequency	32, 33



TYPICAL CHARACTERISTICS





DISTRIBUTION OF TLC27M2 AND TLC27M7 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

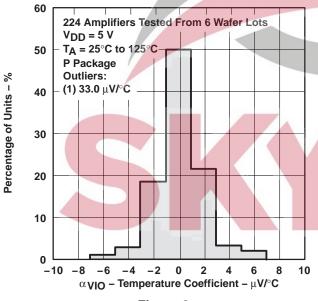


Figure 8

DISTRIBUTION OF TLC27M2 INPUT OFFSET VOLTAGE

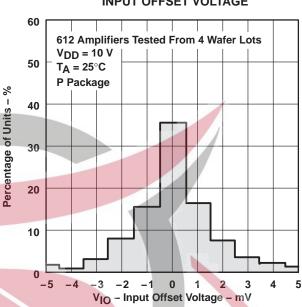


Figure 7

DISTRIBUTION OF TLC27M2 AND TLC27M7 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

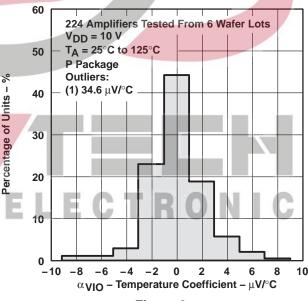
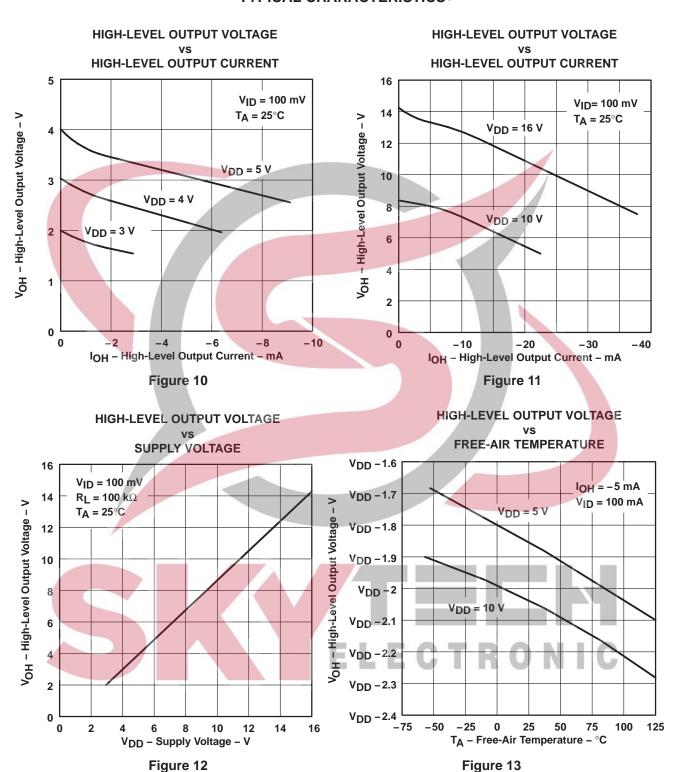
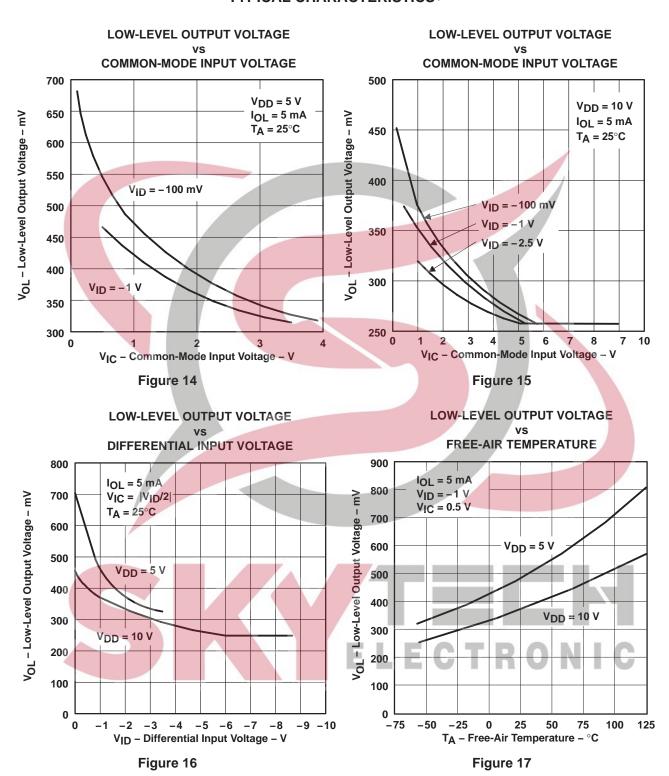


Figure 9



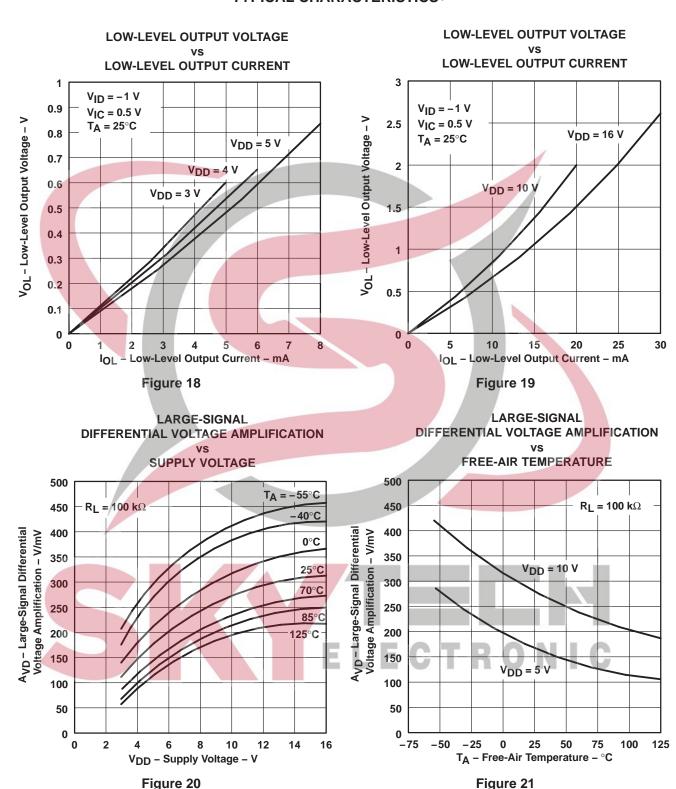
[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





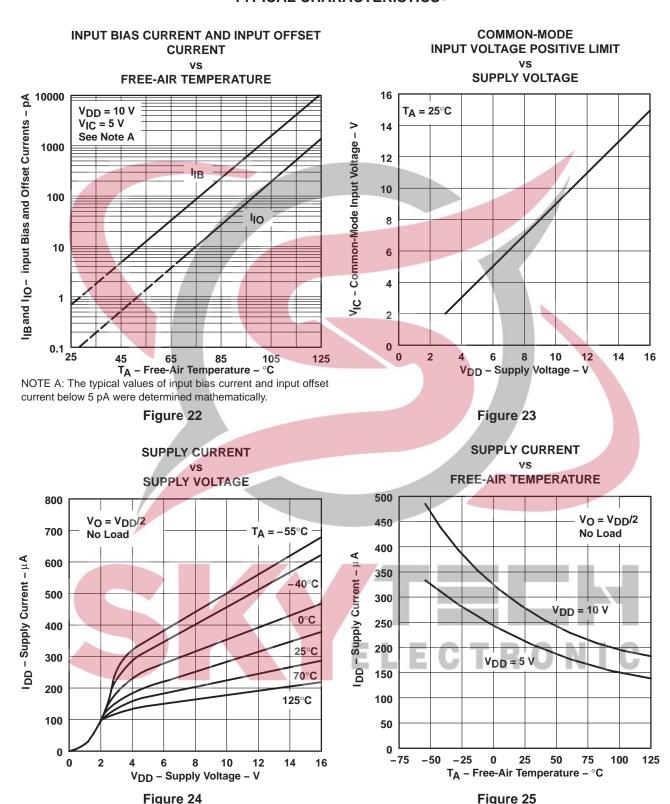
[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





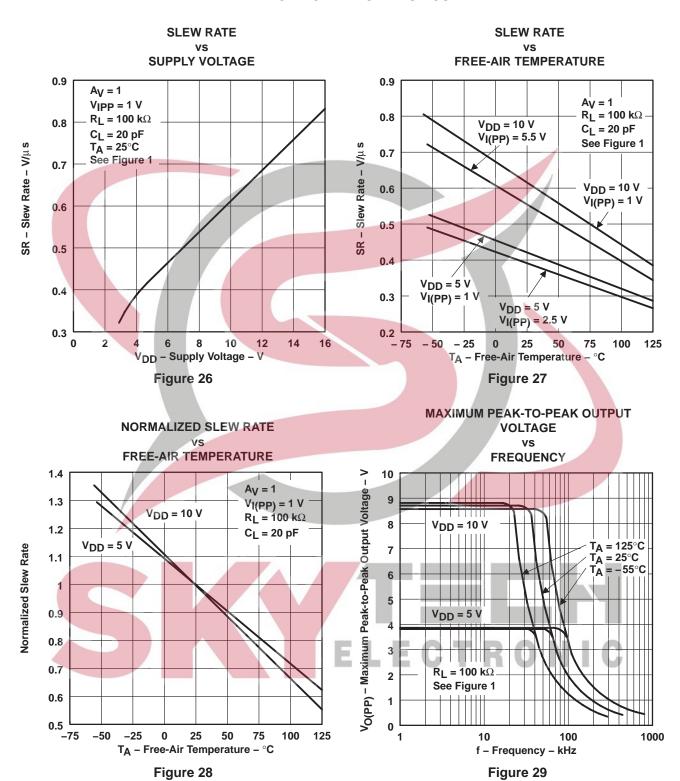
[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





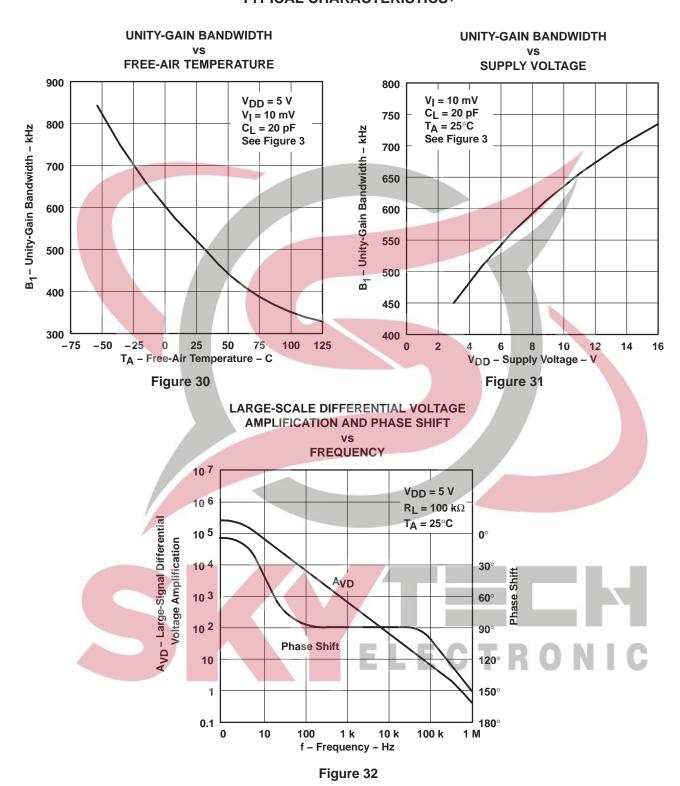
[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





[†]Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



LARGE-SCALE DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT

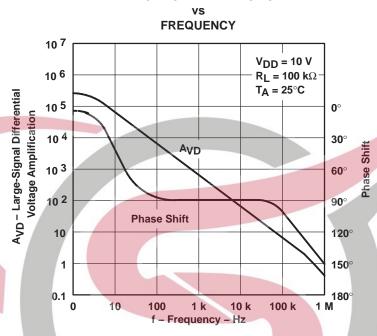
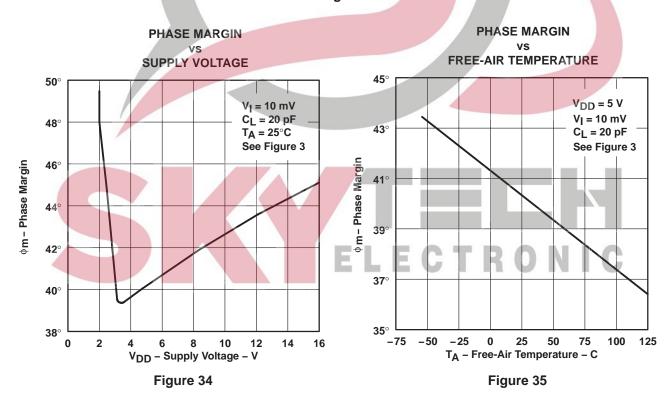


Figure 33



[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

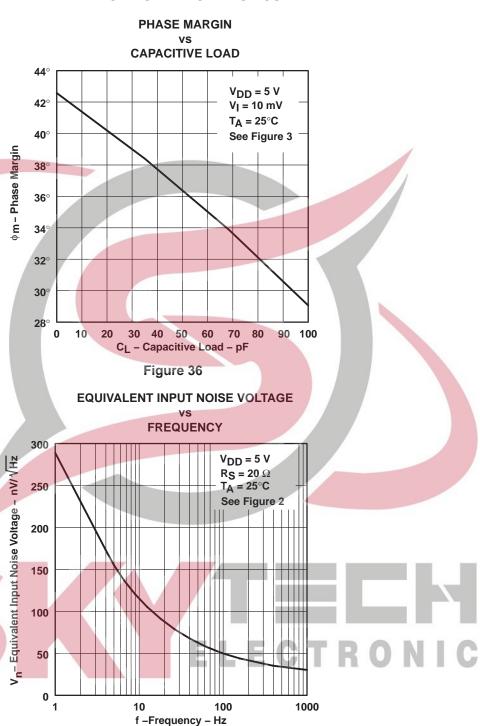


Figure 37

APPLICATION INFORMATION

single-supply operation

While the TLC27M2 and TLC27M7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground, as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M2 and TLC27M7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M2 and TLC27M7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

- 1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
- 2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

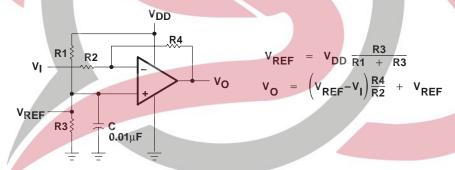
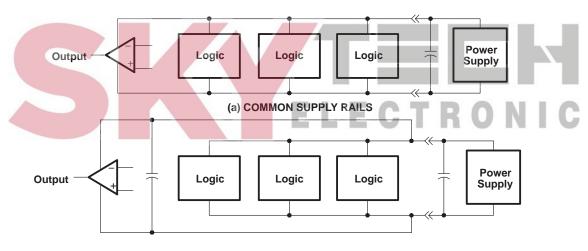


Figure 38. Inverting Amplifier With Voltage Reference



(b) SEPARATE BYPASSED SUPPLY RAILS (preferred)

Figure 39. Common Versus Separate Supply Rails



APPLICATION INFORMATION

input characteristics

The TLC27M2 and TLC27M7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at V_{DD} –1 V at T_A = 25°C and at V_{DD} –1.5 V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M2 and TLC27M7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically 0.1 μ V/month, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M2 and TLC27M7 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M2 and TLC27M7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω , since bipolar devices exhibit greater noise currents.

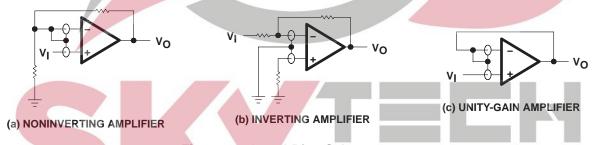


Figure 40. Guard-Ring Schemes

output characteristics

The output stage of the TLC27M2 and TLC27M7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M2 and TLC27M7 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.



APPLICATION INFORMATION

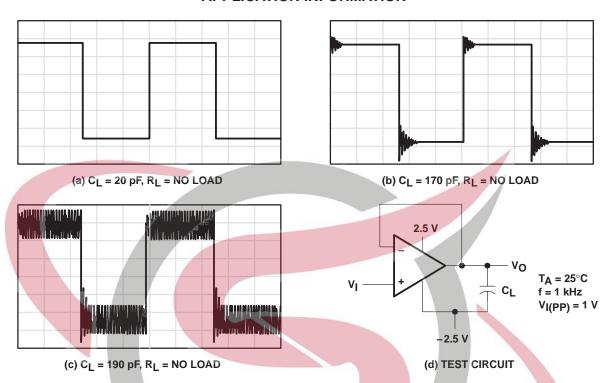


Figure 41. Effect of Capacitive Loads and Test Circuit

output characteristics (continued)

Although the TLC27M2 and TLC27M7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_P) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately $60~\Omega$ and $180~\Omega$, depending on how hard the operational amplifier input is driven. With very low values of R_P, a voltage offset from 0 V at the output occurs. Second, pullup resistor R_P acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.



APPLICATION INFORMATION

output characteristics (continued)

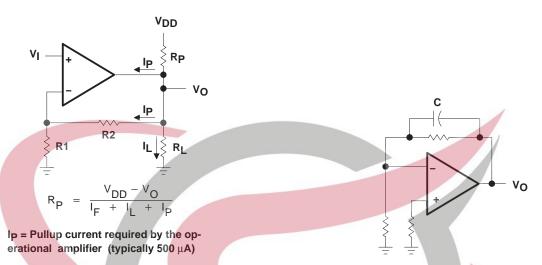


Figure 42. Resistive Pullup to Increase VOH

Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic-discharge protection

The TLC27M2 and TLC27M7 incorporate an internal electrostatic-discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27M2 and TLC27M7 inputs and outputs were designed to withstand -100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.



APPLICATION INFORMATION

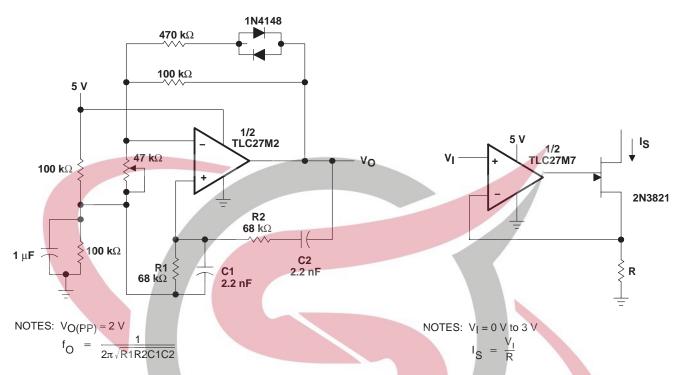


Figure 44. Wien Oscillator

Figure 45. Precision Low-Current Sink

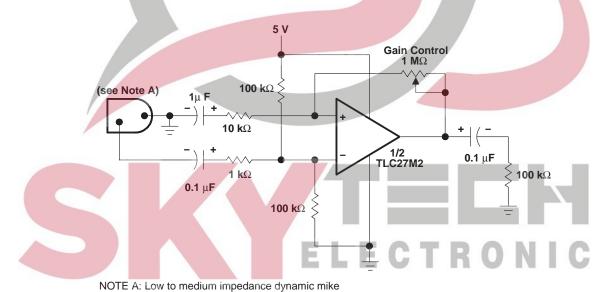


Figure 46. Microphone Preamplifier

APPLICATION INFORMATION

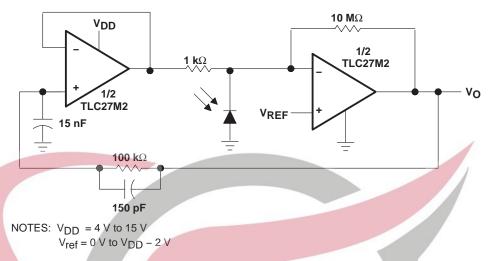


Figure 47. Photo-Diode Amplifier With Ambient Light Rejection

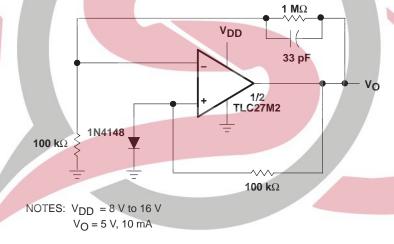
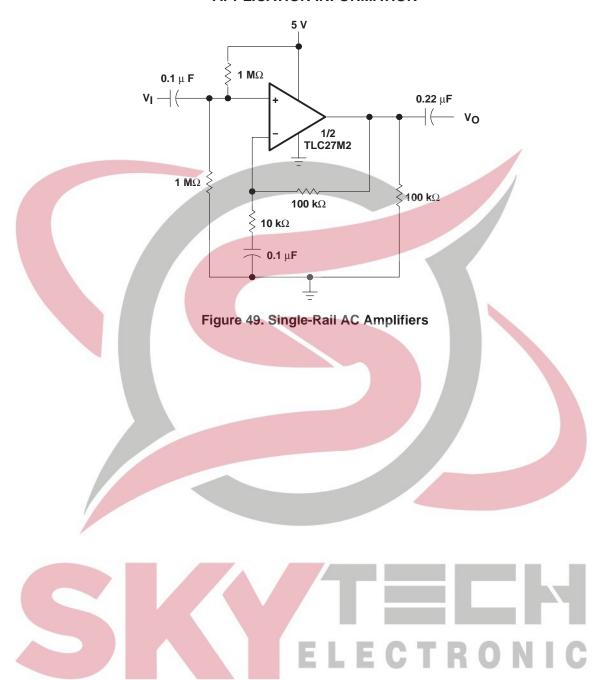


Figure 48. 5-V Low-Power Voltage Regulator



APPLICATION INFORMATION



PACKAGE OPTION ADDENDUM



10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Typ	Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TLC27M2ACD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2AC	Samples
TLC27M2ACDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2AC	Samples
TLC27M2ACDRG4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2AC	Samples
TLC27M2ACP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M2AC	Samples
TLC27M2ACPE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M2AC	Samples
TLC27M2AID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2AI	Samples
TLC27M2AIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2AI	Samples
TLC27M2AIDRG4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2AI	Samples
TLC27M2AIP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M2AI	Samples
TLC27M2BCD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2BC	Samples
TLC27M2BCDG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2BC	Samples
TLC27M2BCDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2BC	Samples
TLC27M2BCP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M2BC	Samples
TLC27M2BID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2BI	Samples
TLC27M2BIDG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2BI	Samples
TLC27M2BIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2BI	Samples
TLC27M2BIP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M2BI	Samples
TLC27M2CD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2C	Samples
TLC27M2CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M2C	Samples
TLC27M2CP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M2CP	Samples

PACKAGE OPTION ADDENDUM



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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TLC27M2CPE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M2CP	Sample
TLC27M2CPS	ACTIVE	SO	PS	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M2	Sample
TLC27M2CPSR	ACTIVE	SO	PS	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M2	Sample
TLC27M2CPW	ACTIVE	TSSOP	PW	8	150	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M2	Sample
TLC27M2CPWR	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M2	Sample
TLC27M2ID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2I	Sample
TLC27M2IDG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2I	Sample
TLC27M2IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M2I	Sample
TLC27M2IP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M2IP	Sample
TLC27M2IPW	ACTIVE	TSSOP	PW	8	150	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	P27M2I	Sample
TLC27M2IPWR	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	P27M2I	Sample
TLC27M2IPWRG4	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M2I	Sample
TLC27M2MD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	27M2M	Sample
TLC27M2MDG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	27M2M	Sample
TLC27M7CD	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M7C	Sample
TLC27M7CDG4	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M7C	Sample
TLC27M7CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M7C	Sample
TLC27M7CP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M7CP	Sample
TLC27M7CPS	ACTIVE	so	PS	8	80	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M7	Sample
TLC27M7CPSR	ACTIVE	so	PS	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M7	Sample
TLC27M7ID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M7I	Sample



10-Dec-2020

Orderable Device	Status (1)	Package Type	Package Drawing		Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TLC27M7IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M7I	Samples
TLC27M7IP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M7IP	Samples
TLC27M7IPE4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M7IP	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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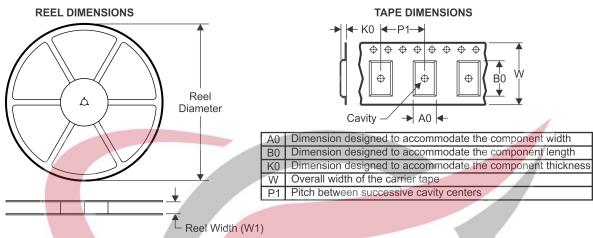
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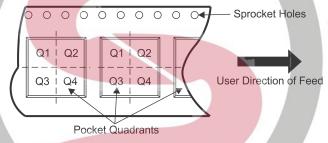
PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION



QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



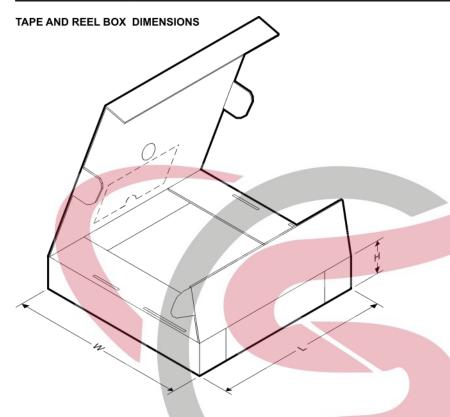
*All dimensions are nominal

All dimensions are nominal												
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC27M2ACDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2BCDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2BIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2CPSR	so	PS	8	2000	330.0	16.4	8.35	6.6	2.5	12.0	16.0	Q1
TLC27M2CPWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TLC27M2IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M2IPWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TLC27M7CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC27M7CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.5	12.0	16.0	Q1
TLC27M7IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



PACKAGE MATERIALS INFORMATION

www.ti.com 30-Dec-2020

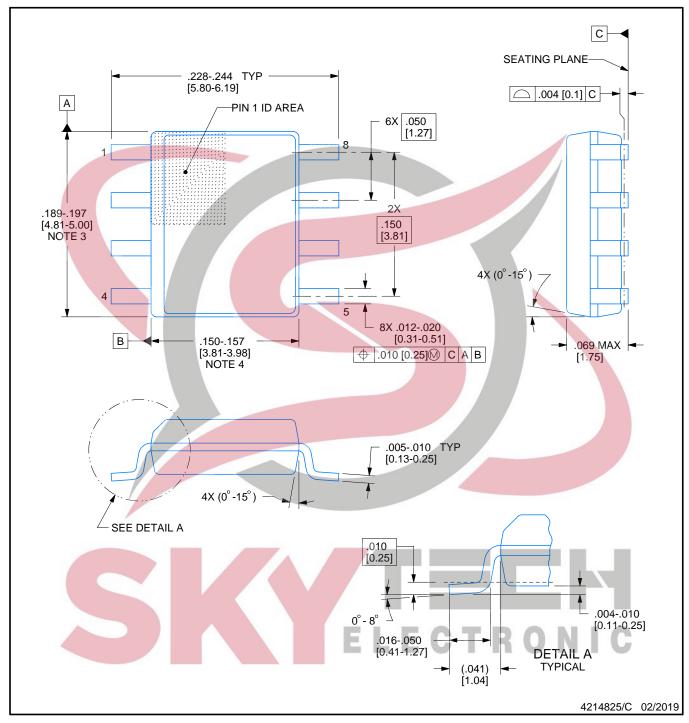


*All dimensions are nominal

7 III dilitionolorio aro nominal							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC27M2ACDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2AIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2BCDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2BIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2CPSR	SO	PS	8	2000	853.0	449.0	35.0
TLC27M2CPWR	TSSOP	PW	8	2000	853.0	449.0	35.0
TLC27M2IDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M2IPWR	TSSOP	PW	8	2000	853.0	449.0	35.0
TLC27M7CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC27M7CPSR	SO	PS	8	2000	853.0	449.0	35.0
TLC27M7IDR	SOIC	D	8	2500	340.5	338.1	20.6

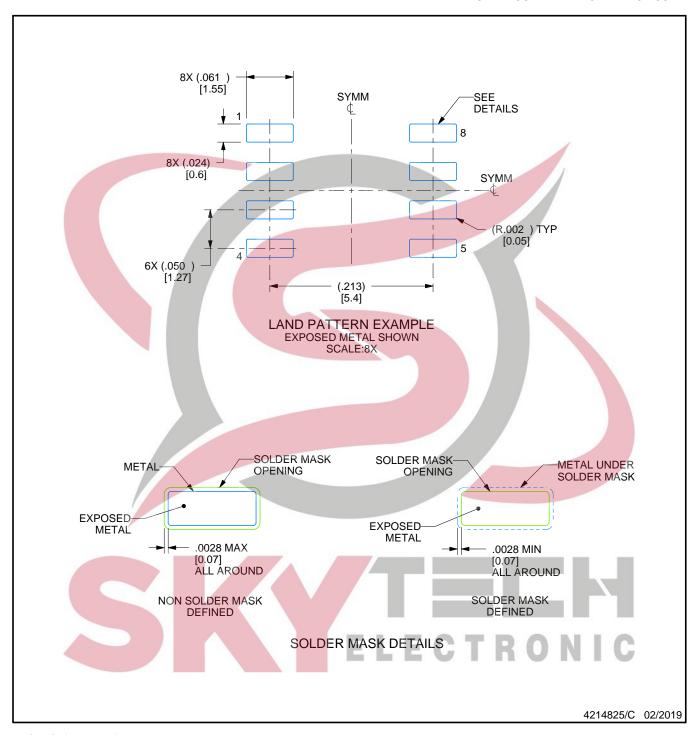


SMALL OUTLINE INTEGRATED CIRCUIT



- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.

SMALL OUTLINE INTEGRATED CIRCUIT

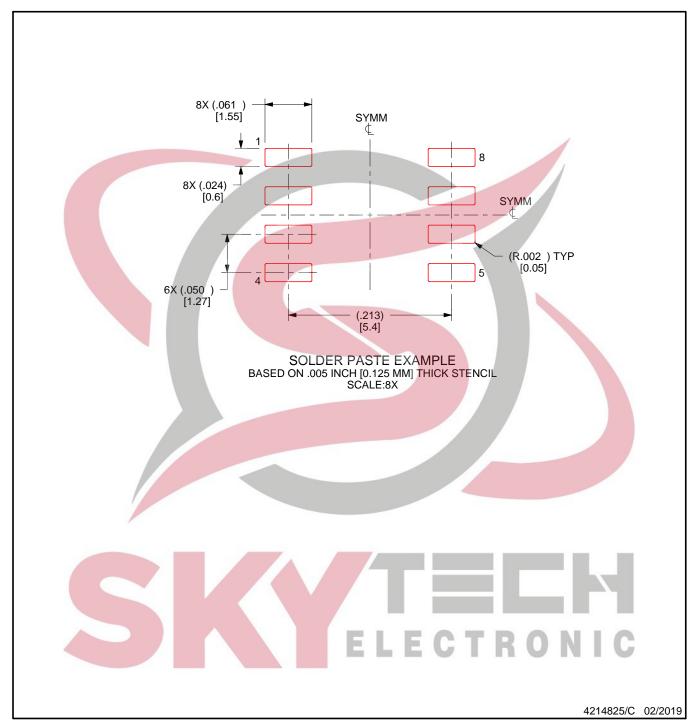


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

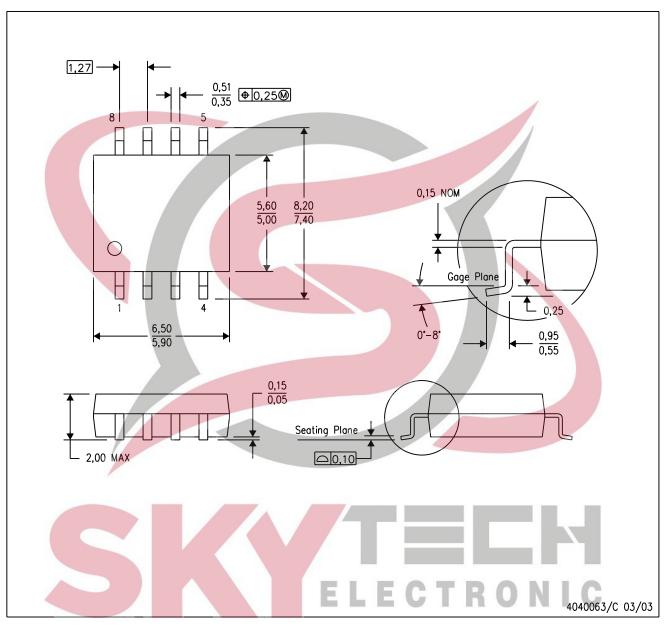
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

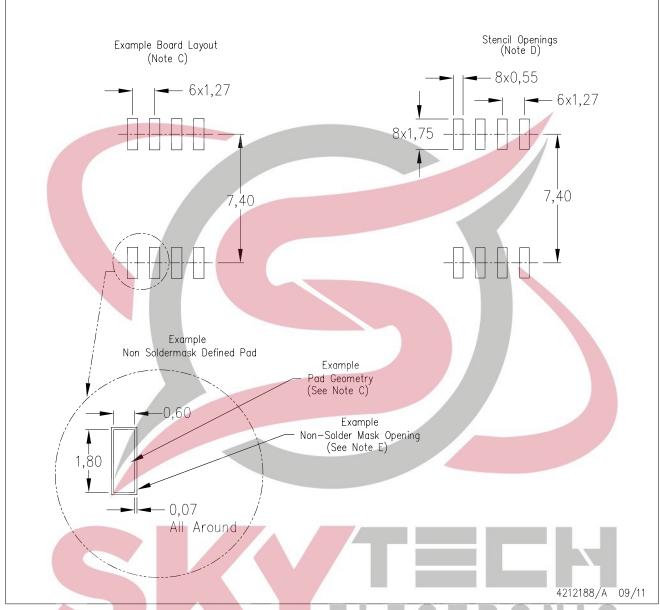
- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



- A. All linear dimensions are in millimeters.
- 3. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

PS (R-PDSO-G8)

PLASTIC SMALL OUTLINE

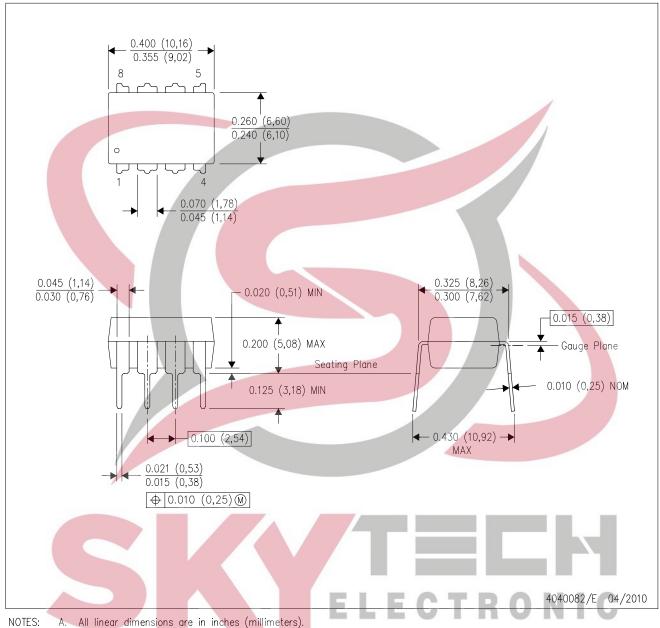


- A. All linear dimensions are in millimeters.
- ELECTRONIC
- B. This drawing is subject to change without notice. C. Publication IPC—7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



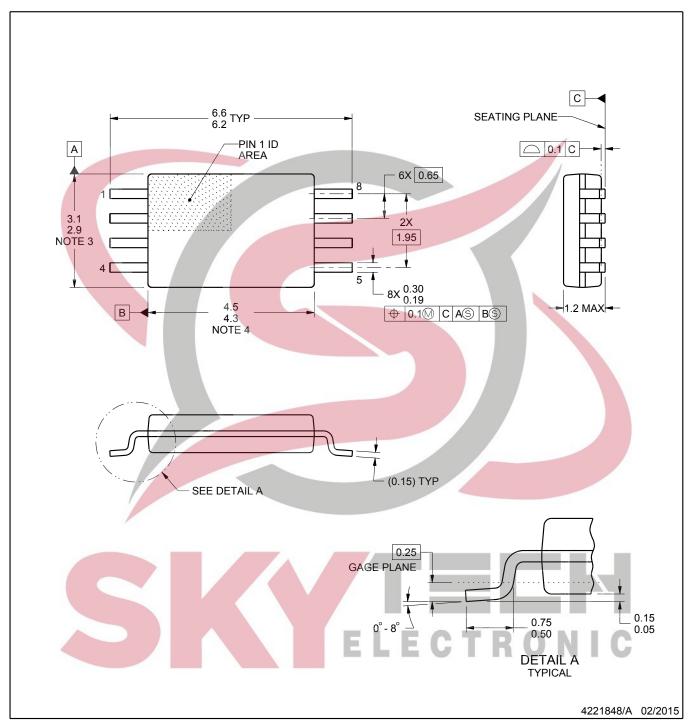
B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-001 variation BA.





SMALL OUTLINE PACKAGE

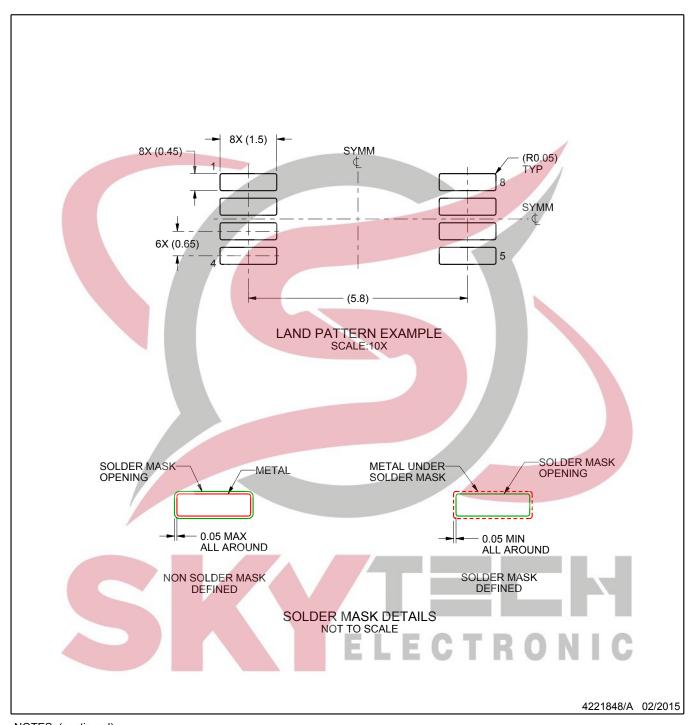


- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153, variation AA.

SMALL OUTLINE PACKAGE

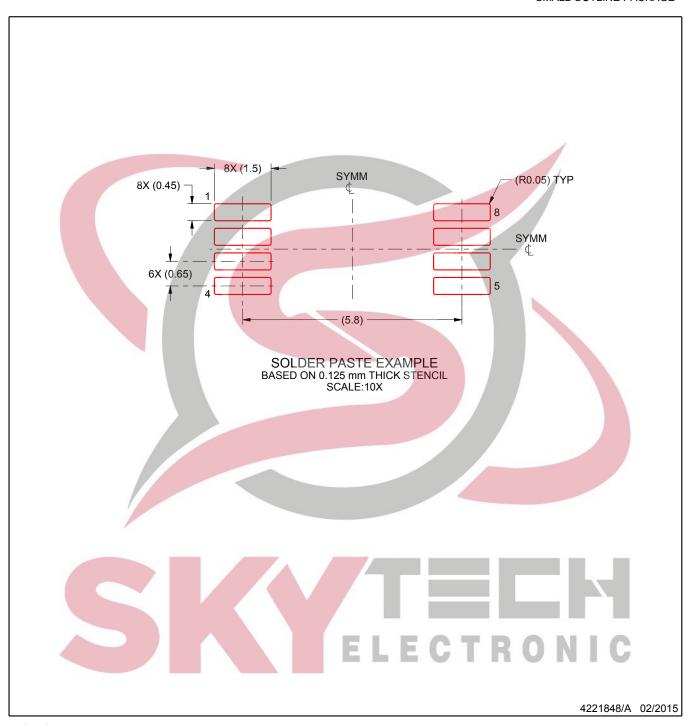


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.

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